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Abstract — This paper presents a low-power and low-voltage 64-kb 8T three-port image memory using a 28-nm FD-SOI process technology. Our proposed SRAM accommodates eight-transistor bitcells comprising one-write/two-read ports and a majority logic circuit to save active energy. The test chip can operate at a supply voltage of 0.46 V and an access time of 140 ns. The energy minimum point is a supply voltage of 0.54 V and an access time of 55 ns (= 18.2 MHz), at which 298 fJ/cycle in a write operation and 650 fJ/cycle in a read operation are achieved with the help of the majority logic; these factor are 87% and 52% smaller than those in a 28-nm FD-SOI 6T SRAM.

Index Terms — Image Memory, Multiport SRAM, 8T, FD-SOI, 28-nm, Majority Logic.

I. INTRODUCTION

Application of image recognition is being extended to various fields such as an automatic driving system, robot vision, and an augmented reality system with improving image resolution. Image resolution enhancement leads the increase of SRAM capacity, area and power consumption because of the image data amount increase. The power consumption in SRAM dissipates 43% of a whole image processor in a 65-nm CMOS process [1]. For wearable devices handling image information, an energy-efficient SRAM will be expected, as shown in Fig. 1.

28-nm Fully-Depleted SOI (FD-SOI) technology is known as promising to provide high speed at low voltage SRAM [2]. 28-nm FD-SOI is implementing fully depleted transistors and an ultra-thin silicon body and BOX layer giving them an excellent electrostatic control. Therefore, 28-nm FD-SOI brings the stable features at low voltage operation. A BOX layer effectively reduces the leakage current to control the electrical flow from a source node to a drain node in a transistor. Moreover, BOX layer reduces the parasitic capacitance between source node and drain node. This feature of 28-nm FD-SOI is allow to make an ultra-low power SRAM [3-7].

In this paper, we design a 28-nm FD-SOI 8T three-port SRAM for a low power image processor and compare it to a 28-nm FD-SOI 6T SRAM. We demonstrate high energy-efficiency of sub-pJ/cycle in the proposed SRAM.

II. PROPOSED THREE-PORT 8T SRAM

A. 8T Three-Port SRAM Cell Design

The circuit schematic of the proposed 8T three-port SRAM is shown in Fig. 2(a); it has a pair of write bitlines and two single-ended read bitlines (one-write/two-read bitcell structure). The proposed SRAM has two pull-up PMOSs (load-PMOS), two pull-down NMOSs (drive-NMOS) and four transfer NMOSs (access-NMOS). In this circuit, M7 and M8 transistors are used as two single-ended read ports. The source nodes of M7 and M8 transistors are connected to node Q, the drain node connected to read bitlines (RBL_A, RBL_B) and the gate node are connected to the read wordlines (RWL_A, RWL_B).

Figs. 2(b) and 2(c) shows the bitcell layout of the proposed 8T three-port SRAM. The size of the SRAM cell is determined by the number of horizontal and vertical wires. In our proposed SRAM, two read ports consists of M7 and M8 transistors are configured as two single-ended read ports have three bitlines and three wordlines. The cell area is 0.56μm² on a logic rule base, which is as small as the dual-port 8T bitcell [8] although the number of ports is increased.

The operating waveforms in the read operation is depicted in Fig. 3. There is no read current flowing through the read bitlines (RBL_A and RBL_B) when the internal node, node Q, is “1”. Maximizing the number of “1”’s at node Q is important to reduce dynamic power in the read operation. In the write cycle, successive data of same values save energy because the proposed SRAM needs not to have a precharge scheme on the write bitlines.
Fig. 2. (a) Schematic, (b) FEOL and (c) BEOL bitcell layouts of the proposed 8T three-port SRAM.

Fig. 3. Waveforms of the proposed 8T three-port SRAM in read operation.

B. SNM of proposed 8T Three-Port SRAM

In the multiport SRAM, simultaneous reads happen at the dual read ports [9]. Fig. 4 shows the variety of read situations in the 1W2R (one-write two-read) three-port SRAM cell. Fig. 4(a) depicts the single port read situation (different two SRAM cells on different row addresses and different column addresses). Fig. 4(b) depicts the simultaneous dual-port reads on same row addresses and column addresses (RWL_A and RBL_B are simultaneously driven in a single cell). In this case, both RBL_A and RBL_B are enabled and double read currents flow from RBL_A and RBL_B to Node Q. Therefore, the read margin (static noise margin) is deteriorated and lowered.

Fig. 5 shows the simulation results of the static noise margin (SNM) at several Vdd voltages from 1.0 V down to 0.4 V. Fig. 5(a) depicts the standard butterfly curves in the single port read situation; the SNM of 171 mV are achieved at 1.0 V, remaining 85% of the SNM in the conventional 6T SRAM [2]. Fig. 5(b) depicts the worst-case butterfly curves in the simultaneous dual-port reads. The SNM is reduces to 101 mV at 1.0 V. The interesting point is that the maximum SNM of 102 mV is observed at 0.8 V.

III. MAJORITY LOGIC

Fig. 6(a) shows a block diagram of the proposed SRAM with the majority logic [10]. Image data reflects luminance information; bright pixels have many “1” data and dark pixels have many “0” data. For the read energy reduction, the dark pixel having many “0”s should be inverted by the majority logic. To maximize the number of “1”s, a majority-logic circuit counts the number of “1”s and decides if input data should be inverted in a write cycle, so that “1”s are in the majority. The inversion information (“1” denotes inversion) is stored in an additional flag bit,
as depicted in Fig. 6(b). In a read cycle, the procedure is reversed. Output data are inverted if a flag bit is true, so that the original data can be read. Note that the majority logic does not reduce write energy because the “1” write energy and the “0” write energy is the same.

The mechanism on the RBL power reduction is shown in Fig. 7. We assume the bit-width of the input data is eight. If the number of “1”s in the input data is five and over, the data is not inverted, and one “0” is stored as flag bit. This means one read current is made by RBL, which is a power overhead. If the number of “1”s in the data is four or less, the data is inverted by majority logic to maximize the number of “1”s, and reduces the read power. When input data have a random pattern, the number of charges/discharges is four out of eight RBLs on average. However, the majority logic reduces this value to 3.27 although the number of RBLs is increased to nine. This indicates that the majority logic statistically saves 18% of an RBL power even if the data are random.

**IV. CHIP IMPLEMENTATION AND MEASUREMENT RESULTS**

We fabricated a 64-kb 8T three-port SRAM macro using a 28-nm FD-SOI process technology. Fig. 8 shows the test chip micrograph. The macro area is 0.058 mm². Fig. 9 shows a measured Shmoo plot of the proposed SRAM macro. We verified that it can operate at a supply voltage of 0.46 V and an access time of 140 ns. At room temperature, the operating point, which achieves the minimum energy per cycle, is a supply voltage of 0.54 V and a cycle time of 55 ns (= 18.2 MHz). Fig. 10 illustrate that the measured leakage and active energies. The write energy is 298 fJ/cycle, which is 78% smaller than that in the 6T SRAM. In the read operation, the “0” and “1” read energies are 1440 fJ/cycle and 148.5 fJ/cycle, respectively. Thus, the energy saving in the “1” read operation is 73%; the read energy improvement is, however, merely 25% on average with no majority logic.

**TABLE I**

<table>
<thead>
<tr>
<th># of “1”s per input data</th>
<th>(0%)</th>
<th>(50%)</th>
<th>(100%)</th>
</tr>
</thead>
<tbody>
<tr>
<td>Power Reduction</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Majority logic with flag bit</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Power overhead by flag bit</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Conventional: (8 bits)</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td># of “0”s in memory cells</td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

**Fig. 7.** A comparison of RBL powers between the conventional and proposed SRAM with majority logic.

**Fig. 9.** Shmoo plot.

**Fig. 11.** Shows the impact of the majority logic on the read energy saving. In the bright Image1, the read energy reduced by 23%, while, in the dark Image1, that reaches a 47% saving; dark image is more appropriate and effective for the majority logic. In this case, the read energy is 650 fJ/cycle. **TABLE I** summarize the characteristics of the test SRAM.
Fig. 10. Measured write energies, read energies, and comparison.

V. SUMMARY

As described in this paper, we presented 8T three-port SRAM for Image processor. The proposed SRAM is comprising one-write/two-read ports and a majority logic circuit to save active energy. We fabricated a 64kb 8T three-port SRAM in a 28-nm FD-SOI process technology. The test chip exhibits 0.46 V operation and an access time of 140 ns. The energy minimum point is a supply voltage of 0.54 V at a frequency of 18.2 MHz, at which 298 fJ/cycle in a write operation and 650 fJ/cycle in a read operation are achieved with the help of the majority logic; these factor are 87% and 52% smaller than those in a 28-nm FD-SOI 6T SRAM.

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