

# on Information and Systems

VOL. E98-D NO. 5 MAY 2015

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A PUBLICATION OF THE INFORMATION AND SYSTEMS SOCIETY



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# PAPER Noise Tolerant Heart Rate Extraction Algorithm Using Short-Term Autocorrelation for Wearable Healthcare Systems

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SUMMARY This report describes a robust method of instantaneous heart rate (IHR) extraction from noisy electrocardiogram (ECG) signals. Generally, R-waves are extracted from ECG using a threshold to calculate the IHR from the interval of R-waves. However, noise increases the incidence of misdetection and false detection in wearable healthcare systems because the power consumption and electrode distance are limited to reduce the size and weight. To prevent incorrect detection, we propose a short-time autocorrelation (STAC) technique. The proposed method extracts the IHR by determining the search window shift length which maximizes the correlation coefficient between the template window and the search window. It uses the similarity of the QRS complex waveform beat-by-beat. Therefore, it has no threshold calculation process. Furthermore, it is robust against noisy environments. The proposed method was evaluated using MIT-BIH arrhythmia and noise stress test databases. Simulation results show that the proposed method achieves a state-of-the-art success rate of IHR extraction in a noise stress test using a muscle artifact and a motion artifact.

*key words:* autocorrelation, biomedical signal processing, electrocardiography, heart rate extraction, noise tolerance

## 1. Introduction

Mobile and wearable healthcare devices are expected to play an increasingly prominent role in health provision due to the advent of aging societies around the world [1]. In particular, biosignal measurements during daily life at home are important to prevent lifestyle diseases, which are expected to raise the number of patients and elderly people requiring nursing care.

This report describes a noise-tolerant instantaneous heart rate (IHR) extraction algorithm from noisy ECG signals. The IHR is useful for heart disease detection, heart rate variation analysis [2], and exercise intensity estimation [3].

Key factors affecting wearable system usability are miniaturization and weight reduction. A wearable and wireless ECG telemetry system [4], [5] and single-chip ECG monitoring system LSIs [6]–[8] have been developed. However, the wearable system is sensitive to noise because of strict limitations on power consumption and electrode distance (see Fig. 1). The signal-to-noise ratio (SNR) of ECG signals can be degraded, especially if a subject is not at rest.

In general, sophisticated analog front-end circuits are necessary to prevent SNR degradation. The analog

Manuscript received May 17, 2014.

<sup>†</sup>The authors are with the Graduate School of System Informatics, Kobe University, Kobe-shi, 657–8501 Japan. front-ends of ECG monitoring systems mainly comprise amplifiers, analog filters, and an analog-to-digital converter (ADC). Unfortunately, analog circuits have a large circuit area and high power consumption. Battery mass and power consumption must be reduced because the battery mass dominates wearable systems. The amplifier presents a tradeoff between power consumption and performance (e.g., gain, phase characteristic, common mode rejection ratio). Moreover, the analog filter in an ECG monitor has a large RC time constant because the ECG signal frequency range is low (f < 1 kHz). For those reasons, it is difficult to use high-performance amplifiers and analog filters that have a high quality factor.

Ultra-low-power ADCs, which have sub-microwatt power consumption and a limited sample rate, have been developed for biomedical applications [9], [10]. Furthermore, according to Moore's law, the power of digital components increases with the progress of process technology. In contrast, the power consumption of analog circuits will not decrease similarly. Therefore, as illustrated in Fig. 2, our approach using digital signal processing aims to reduce the performance requirements of analog components and to minimize the system's overall power consumption.

A preliminary version of this work has been reported in the literature [11]. This paper presents an extended version of the algorithm and additional details of performance evaluation results. Section 2 of this report describes conventional



Fig. 1 Constraints of a wearable healthcare system.

Manuscript revised November 16, 2014.

Manuscript publicized January 26, 2015.

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DOI: 10.1587/transinf.2014EDP7161



Fig. 2 Block diagram of ECG sensing circuits and target of this work.

heart rate extraction techniques. An improved heart rate extraction scheme is proposed in Sect. 3. Section 4 presents some simulation results including noise stress tests. Finally, conclusions are presented in Sect. 5.

### 2. Heart Rate Extraction Techniques

Recently, various algorithms have been proposed to improve heart rate extraction accuracy and reliability.

Extracting R-waves using threshold determination is a widely used approach for IHR extraction from ECG. The Pan–Tompkins (PT) algorithm [12], which is commonly used for beat detection, uses band-pass filtering, differentiation, squaring, and moving window integration. Periodically, the threshold is adjusted automatically using QRS morphology and the heart rate.

The SQRS [13] and WQRS [14] algorithms, which have been published in PhysioNet, can respectively detect QRS based on ECG slope and length transform. The SQRS uses band pass filtering for noise reduction, which uses only the integer coefficient. The WQRS also uses a low-pass filter to remove baseline wander.

The Discrete Wavelet Transform (DWT) [7], [15], [16] uses a wavelet transform with quadratic spline wavelet (QSW). The threshold is calculated using the root mean square value of the wavelet transform. This algorithm has been used in robust ECG monitoring LSIs [7], [17], [18]. The QSW requires few calculations and low hardware costs because it can be implemented using only adders and shift operators.

The Quad Level Vector (QLV) algorithm [19] is used in dedicated hardware for ECG monitoring LSI [6], [20]. The QLV is generated using DWT and the adaptive threshold. Then, the threshold is ascertained from the maximum mean deviation (MD) of prior heartbeats.

The Continuous Wavelet Transform (CWT) algorithm [21]–[23] uses a Mexican hat wavelet in the frequency interval of 15–18 Hz. The R-peak can be extracted using the adaptive threshold, which is calculated using the modulus maxima of the CWT. This algorithm was also implemented in an earlier study [6].

Figure 3 presents the heart beat detection accuracy of conventional threshold based algorithms from the 1980s [12], [15], [16], [22], [24]–[46]. As depicted in Fig. 3, no significant difference was found in the accuracy. However, these results are evaluated using only clean ECG (MIT-BIT open ECG database record #100 [47]). As depicted in Fig. 4, noise from various sources increases misdetection and false detection in the wearable healthcare system. Figure 5 (a) presents frequency characteristics of the PT, SQRS,



Fig. 3 Accuracy comparison of recent heart beat detection methods with MIT-BIH record #100.



Fig. 4 Noise problem with threshold based R-peak detection.



**Fig.5** (a) Frequency characteristics of filters with 128 Hz sampling rate. (b) Waveform of ECG signals with noise.

and DWT with a 128 Hz sampling rate. Figure 5 (b) depicts the ECG and well-known noise waveforms. A baseline wander and a hum noise can be removed easily using digital filters. However, unfortunately, the frequency ranges of the muscle artifact and electrode motion artifacts are similar to those of the desired ECG signals.

Therefore, this work was undertaken for noise tolerance improvement. Threshold-based algorithms are classifiable using preprocessing and QRS detection, as presented in Table 1. Our proposed method, as described in Sect. 3,

	Preprocessing (filter technique)	QRS detection (threshold)
[12]	Bandpass filter, derivative, squaring, and moving window integrator	Signal peak and noise peak
[13]	Bandpass filter	Slope criterion
[14]	Lowpass filter and curve length transform	QRS amplitude
[16]	DWT	Root Mean Square (RMS)
[19]	Discrete wavelet transform (DWT)	Mean deviation (MD)
[22]	Continuaous wavelet transform (CWT)	Modulus maxima

 Table 1
 Preprocessing and QRS detection in conventional methods.

can replace the threshold-based method. In other words, the proposed method can be combined with any other preprocessing filter technique shown in Table 1.

# 3. Short-Term Autocorrelation

In this work, we propose a short-term autocorrelation (STAC) technique for IHR extraction. Autocorrelation [48], [49] and template matching [50] use the similarity of QRS complex waveforms and have no threshold calculation process. Autocorrelation has been used in non-invasive monitoring systems. However, the method necessitates numerous computations because it calculates the average heart rate over a long duration (30 s). In this work, we extend it for IHR extraction by minimizing the window length because conventional works can only extract the average heart rate from data of a long duration. To achieve accurate heart rate variability analysis, the IHR of every second is required.

Figure 6 portrays IHR extraction using STAC. As depicted in Fig. 6 and (1–7), the recent interval of R-waves at time  $t_n$  (*RR*[n]) is obtained as a window shift length ( $T_{shift}$ ) that maximizes the correlation coefficient between the template window and the search window (*CC*[n]). The IHR at  $t_n$  (*IHR*[n]) is calculable using *RR*[n], as shown in (6).

$$RR[n] = \arg_{T_{\text{shift}}} \max_{RR_{\min} \leq T_{\text{shift}} \leq RR_{\max}} \{CC[n]\}$$
(1)  
$$CC[n] = W_{c}(T_{\text{shift}}) \cdot \sum_{i=0}^{L_{w}[n]} W_{w}(i) \cdot d[t_{n} - i]$$

$$\cdot d[t_{\rm n} - i - T_{\rm shift}] \tag{2}$$

$$W_{\rm c}(T_{\rm shift}) = \max\left\{1 - \frac{1}{4} \cdot \left\lfloor \frac{T_{\rm shift} - RR_{\rm min}}{RR_{\rm min}} \right\rfloor, \frac{1}{4}\right\}$$
(3)

$$W_{w}(i) = \max\left[1 - \frac{1}{8} \cdot \left\lfloor \frac{t}{RR_{min}} \right\rfloor, \frac{1}{8}\right]$$
(4)  
$$L_{w}[\mathbf{n}]$$

$$= \begin{cases} RR[n-1] \times (1.2)^3 & (RR_{\min} \le RR[n-1] < 0.316) \\ RR[n-1] \times (1.2)^2 & (0.316 \le RR[n-1] < 0.695) \\ RR[n-1] \times (1.2)^1 & (0.695 \le RR[n-1] \le RR_{\max}) \end{cases}$$
(5)

$$IHR[n] = \frac{60}{RR[n]} \tag{6}$$

$$t_{\rm n} = t_{\rm n-1} + 1 \tag{7}$$

In the equations presented above,  $W_c$  and  $W_w$  denote weight



Fig. 6 IHR extraction using short-term autocorrelation (STAC).



**Fig.7** Weight coefficient *W*<sub>c</sub> to suppress old R-peak misdetection.



**Fig. 8** Weight coefficient *W*<sub>w</sub> to improve R-peak detection accuracy.

coefficients. As depicted in Fig. 7,  $W_c$  contributes to the choice of the recent peak of the correlation coefficient if two or more R-peaks exist in the shift range of  $T_{\text{shift}}$ ;  $W_w$  contributes to the choice of the accurate peak of the correlation coefficient if two or more R-peaks exist in the template of search window because the waveform of the correlation coefficient has a bimodal peak in such a case (see Fig. 8).

The shift range of  $T_{\text{shift}}$  is decided by maximum and minimum values of RR[n] ( $RR_{\text{max}}$  and  $RR_{\text{min}}$ ). In this work,  $RR_{\text{max}}$  and  $RR_{\text{min}}$  are set respectively as 0.25 s and 1.5 s because the heart rate of a healthy person is 40 bpm–240 bpm in general.

The  $L_w$  in (1) and (3) denotes the window length, which is updated according to the estimated value of RR[n] from RR[n - 1], as shown in (5). The initial value of  $L_w$  is set to 1.5 s in this work.  $L_w$  should be set to include one or more beats in the template and search windows. If RR[n-1]is smaller than 0.316 s, then three R-waves exist between  $t_n - I$  and  $t_n$  at maximum. Then  $L_w$  is set to  $RR[n-1] \times (1.2)^3$  according to the maximum rate of the beat-to-beat variation, which is generally 20% in a healthy person [51]. Similarly, if the RR[n-1] is 0.3160.695 s and larger than 0.695 s, then the respective  $L_w$  settings are  $RR[n-1] \times (1.2)^2$  and  $RR[n-1] \times (1.2)$ . This  $L_w$  optimization (5) contributes to reduction of the computational amount and to improvement of the IHR estimation accuracy.

Unfortunately, the computational cost of the proposed method is about one hundred times as great as that for general threshold methods. Nevertheless, this method can be implemented in the digital domain. We estimate that the power consumption of the proposed method is one-tenth that of the analog portion, which includes an instrumental amplifier, analog filter, and ADC. Furthermore, the power consumption of the analog portion will be reduced using the proposed method because it has higher noise tolerance. Therefore, the total power consumption of the wearable monitor can be reduced.

# 4. Performance Evaluation

To evaluate the noise tolerance of heart rate extraction algorithms, we implemented the proposed STAC algorithm and conventional algorithms using MATLAB. The objective of the proposed method is IHR extraction every second. Therefore, the proposed method cannot detect all heart beats in the ECG if the interval of R-waves is less than 1 s. To compare the IHR extraction success rate every second, the IHR of the conventional threshold-based method is calculated from the recent interval of R-waves at evaluation time  $t_n$  in (7).

#### 4.1 IHR Extraction Success Rate with Clean ECG

First, we investigated the success rate of heart rate extraction using 48 records from the MIT-BIH arrhythmia database [47]. As described above, the proposed STAC can be combined with any other preprocessing filter technique shown in Table 1. The DWT [16] and CWT [22] are implemented as filters in this simulation because the DWT can be realized by simple implementation in hardware and because the CWT is the most noise-tolerant filter technique for ECG.

The threshold of conventional methods is calculated as explained below: In Pan-Tompkins, the threshold is calculated using (ECG<sub>max</sub>/8+(Prev<sub>TH</sub>-Prev<sub>TH</sub>/8)). Here, Prev<sub>TH</sub> and ECG<sub>max</sub> respectively denote the previous value of the threshold and the maximum value of ECG in 2 s. In SQRS, the threshold is calculated using ((Prev<sub>TH</sub>+ECGABS<sub>max</sub>/4-Prev<sub>TH</sub>)/8). Here, ECGABS<sub>max</sub> denotes the maximum of the absolute value of ECG in 2 s search range. In WQRS, the average value of ECG in 8 s is used first. Subsequently, the threshold is updated by adding (ECG<sub>max</sub> - Prev<sub>TH</sub>/3). In QLV, the ECG is divided into blocks of 0.04 s duration. The threshold is calculated using half of the mean deviation (MD) value of eight blocks that have a larger MD value than the threshold. In DWT, the threshold is calculated using the eighth of the root mean square value of ECG in prior 2 s.

 
 Table 2
 IHR extraction success rate of proposed STAC with DWT and CWT filters for MIT-BIH waveforms.

	Success rate [%]			Success rate [%]		
Record	w/ DWT	w/ CWT	Record	w/ DWT	w/ CWT	
100	99.44	99.89	202	94.50	96.33	
101	99.61	99.67	203	63.00	77.50	
102	93.28	94.84	205	97.50	99.56	
103	99.94	99.94	207	85.62	89.45	
104	93.11	93.56	208	52.97	47.08	
105	96.00	97.72	209	99.28	99.78	
106	79.66	82.32	210	85.34	90.89	
107	94.17	98.72	212	99.89	100.00	
108	80.34	81.51	213	93.06	95.45	
109	98.22	99.06	214	89.28	92.95	
111	95.72	96.50	215	96.78	96.78	
112	100.00	100.00	217	89.72	90.44	
113	98.72	98.45	219	91.83	94.22	
114	93.83	99.06	220	98.28	98.89	
115	99.83	99.17	221	84.40	85.84	
116	96.45	97.17	222	88.62	88.73	
117	94.84	94.28	223	91.50	95.28	
118	98.00	98.67	228	82.83	70.56	
119	67.96	70.41	230	99.94	100.00	
121	99.72	99.94	231	99.17	99.56	
122	100.00	100.00	232	71.02	72.65	
123	99.61	99.22	233	57.13	88.34	
124	97.94	98.67	234	99.78	<u>99.</u> 72	
200	42.64	88.12	Average	80.72	02.48	
201	75.94	82.00	Average	09.72	92.40	

 Table 3
 Relation between success rate and type of arrhythmia in 48 records of MIT-BIH.

	# of beats	Success rate [%]				
Type of arrhythmia		Conventional		Proposal		
		DWT	CWT	w/ DWT	w/ CWT	
Normal beat	52927	93.18	96.16	97.52	97.88	
Left bundle branch block beat	6372	82.42	97.65	96.39	97.65	
Right bundle branch block beat	6366	74.55	96.26	94.53	95.62	
Atrial premature beat	2588	92.31	93.59	82.65	83.81	
Aberrated atrial premature beat	258	58.53	33.72	41.86	38.76	
Nodal (junctional) premature beat	64	68.75	95.31	85.94	87.50	
Premature ventricular contraction	9847	62.96	63.18	41.85	61.10	
Fusion of ventricular and normal beat	715	84.90	92.31	90.49	94.97	
Atrial escape beat	21	100.00	90.48	90.48	90.48	
Nodal (junctional) escape beat	252	82.94	92.46	83.33	84.52	
Ventricular escape beat	108	87.96	92.59	88.89	93.52	
Paced beat	5880	71.04	92.45	94.63	96.02	
Fusion of paced and normal beat	854	56.91	76.93	85.48	86.30	
Unclassifiable beat	35	8.57	57.14	51.43	77.14	
Ventricular flutter wave	147	14.29	26.53	42.86	34.69	

In CWT, the threshold is calculated using 30% of the maximum value of ECG in the prior 2 s.

Table 2 presents the simulation results. Compared with the conventional algorithms depicted in Fig. 3, no significant difference was found in the success rate obtained with record #100. However, for two reasons, the success rate was degraded for several records (e.g. #119, #208, and #233). The first reason for the performance degradation is a certain type of arrhythmia, which has irregular heart beat waveform (e.g. premature ventricular contraction). As Table 3 shows, although the proposed method shows equivalent or better



Fig. 9 Noise stress test using MIT-BIH record #100 with motion artifact.



Fig. 10 Noise stress test using MIT-BIH record #100 with muscle artifact.

performance in most cases, it is degraded by such types of arrhythmia because the proposed algorithm uses similarity of the QRS waveform. Therefore, it is difficult to detect a sudden change in the QRS waveform.

The second reason is heart rate variation. Although we assume that the maximum rate of the beat-to-beat variation is 20%, several records include 30% or greater variation. Although this problem can be solvable through parameter tuning, a tradeoff exists between the success rate and the computational amount.

#### 4.2 Performance Comparison in Noise Stress Tests

Next, we evaluated the noise tolerance using the MIT-BIH noise stress test database [52]. Figures 9–12 show the relation between the noise intensity and the success rate of IHR extraction. The MIT-BIH records #100 and #122 are used to evaluate the effects of noise contamination and to eliminate the effects of arrhythmia because these records include few arrhythmia beats. A muscle artifact and motion artifact records are used because these noises have critical frequency characteristics, as presented in Fig. 5. Then, the signal-to-noise ratio (SNR) is defined as shown below.



Fig.11 Noise stress test using MIT-BIH record #122 with motion artifact.



Fig. 12 Noise stress test using MIT-BIH record #122 with muscle artifact.



Fig. 13 Noise stress test using MIT-BIH record #100 with motion artifact. Conventional detection methods are combined with CWT filter.

$$SNR = 10\log\frac{S}{N \times a^2} \tag{8}$$

Here, S, N, and a respectively denote the signal power, frequency-weighted noise power, and scale factor.

Simulation results show that the proposed STAC can improve noise tolerance in both cases of combination: with



**Fig. 14** Noise stress test using MIT-BIH record #100 with muscle artifact. Conventional detection methods are combined with CWT filter.



Fig. 15 Noise stress test using MIT-BIH record #122 with motion artifact. Conventional detection methods are combined with CWT filter.

the DWT filter and with the CWT filter. The combination of the STAC and CWT filter achieves the top of noise tolerance with both the muscle artifact and the motion artifact.

Furthermore, in Figs. 13–16, the proposed method is compared to the combination of the CWT filter with conventional QRS detection (threshold) methods, as shown in Table 1. Results show that the success rate of conventional detection methods with CWT filter is improved from original method. This result demonstrates that the CWT filter itself can improve the performance. Note that the combination of the CWT filter and the proposed STAC method still shows better performance in most cases compared with other combinations. Therefore, both the CWT filter and the proposed STAC method contribute synergistically to improved performance.

# 4.3 Required Resolution of ECG for Hardware Implementation

Finally, we evaluate the required resolution of the ECG signal. The computational amount and the hardware overhead



**Fig. 16** Noise stress test using MIT-BIH record #122 with muscle artifact. Conventional detection methods are combined with CWT filter.



**Fig. 17** Bit width of ECG signal versus average success rate of all MIT-BIH waveforms in Table 2 with and without 9-dB motion artifact. The sampling rate is set to 128 samples/s.



**Fig. 18** Sampling rate of ECG signal versus average success rate of all MIT-BIH waveforms in Table 2 with and without 9-dB motion artifact. The bit resolution is set to eight bits.

of IHR extraction should be minimized because the battery capacity is strictly limited in our target application. The bit width and the sampling rate of ECG signal directly affect the overhead.

Figure 17 presents simulation results of the average IHR extraction success rate from 48 records in Table 2 with 4-bit to 11-bit width. The sampling rate of the ECG signal is set to 128 samples/s in this simulation. Results show that

the success rate is degraded when the bit width is less than six in both clean and noisy conditions.

Figure 18 portrays the effects of sampling rate differences. The bit width is fixed to eight bits. Other simulation conditions are presented in Fig. 17. Simulation results show that a 128 samples/s sampling rate is needed for extraction of IHR without degradation.

#### 5. Conclusion

As this report has described, we proposed a noise-tolerant IHR extraction algorithm using short-term autocorrelation (STAC). We limited the window length to 1.5 s according to the heart rate of a healthy person, which is 40-240 bpm in general. The window length should be longer than the maximum R2R interval because at least one heart beat should be presented in the window. To realize accurate heart rate extraction using the minimum window length, there are two improved points against the previous autocorrelation algorithm. First, two weight coefficients are introduced to minimize incorrect peak detection both in the window and in the search range. Next, we combined the STAC with DWT and CWT filters because the window length reduction causes noise tolerance degradation. Therefore, we can achieve both noise tolerance improvement and computational reduction. Simulation results show that the IHR extraction success rate of the proposed STAC with the CWT filter is 99.89% for MIT-BIH record #100 and 92.48% on average for 48 records of MIT-BIH. In the noise stress test, the proposed method achieves state-of-the-art noise tolerance both with the muscle artifact and the motion artifact. The required resolution of ECG signal is also evaluated. The proposed method requires seven-bit width and a 128 samples/s sampling rate for ECG signals to extract the IHR without success rate degradation.

#### Acknowledgments

This research was partially supported by the Ministry of Economy, Trade and Industry (METI) and the New Energy an Industrial Technology Development Organization (NEDO) and a grant from the Tateishi Science and Technology Foundation.

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